

Green-Mode PWM Controller

REV: 01a

General Description

The LD7550-B is a low cost, low startup current, current mode PWM controller with green-mode power-saving operation. The integrated functions such as the leading-edge blanking of the current sensing, internal slope compensation and the small SOT-26 package provide the users a high efficiency, low external component counts, and low cost solution for AC/DC power applications.

Compared with LD7550, LD7550-B has different parameters or functions on the following characteristics ---

- Enlarger the hysteresis range of UVLO by lower down the UVLO(off) from 11.4V to 10V.
- Add OVP (Over Voltage Protection) function on Vcc Pin.
- Shorten the OCP delay time to 100nS to reduce the difference of OCP trip level under high line & low line.

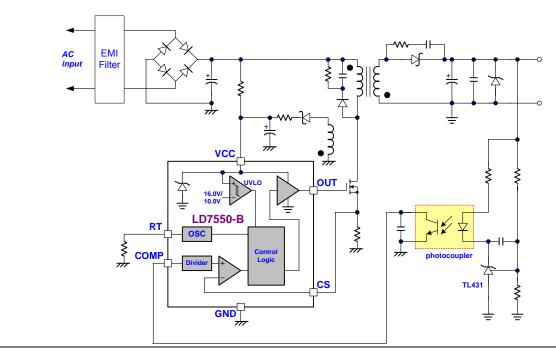
Features

- High-Voltage CMOS Process with Excellent ESD protection
- Very Low Startup Current (<20μA)
- Under Voltage Lockout (UVLO)
- Current Mode Control with Cycle-by-Cycle Peak Current Limiting
- Leading-Edge Blanking on CS Pin
- Programmable Switching Frequency
- Internal Slope Compensation
- Proprietary Green-Mode Control for Power Saving
- Non-audible-noise Green Mode Control
- 300mA Driving Capability
- OVP (Over Voltage Protection) on Vcc Pin

Applications

- Switching AC/DC Adaptor and Battery Charger
- Open Frame Switching Power Supply
- 384X Replacement

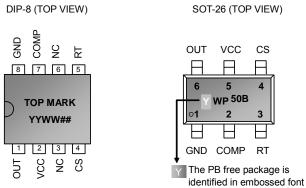
Typical Application



1



Pin Configuration



YY, Y: Year code (D: 2004, E: 2005....)

WW, W: Week code P : LD75..

(Product family code)
: Production code

Ordering Information

Part number	Package	TOP MARK	Shipping
LD7550-B BL	SOT-26	YWP/50B	3000 /tape & reel
LD7550-B BN	DIP-8	LD7550BBN	3600 /tube /Carton

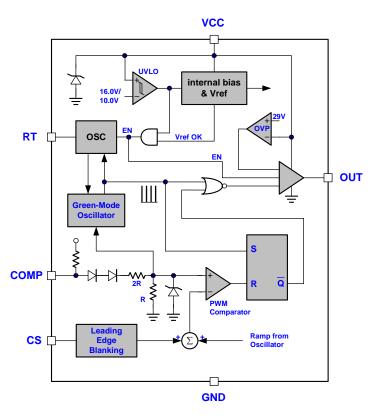
The LD7550B is ROHS compliant.

Pin Descriptions

PIN SOT-26	PIN DIP-8	NAME	FUNCTION
1	8	GND	Ground
2	7	COMP	Voltage feedback pin (same as the COMP pin in UC384X), By connecting a photo-coupler to close the control loop and achieve the regulation.
3	5	RT	This pin is to program the switching frequency. By connecting a resistor to ground to set the switching frequency.
4	4	cs	Current sense pin, connect to sense the MOSFET current
5	2	VCC	Supply voltage pin
6	1	OUT	Gate drive output to drive the external MOSFET
	3, 6	NC	Unconnected pin



Block Diagram



Absolute Maximum Ratings

Supply Voltage VCC	36V
COMP, RT, CS_	-0.3 ~7V
Junction Temperature	150°C
Operating Ambient Temperature	-40°C to 85°C
Storage Temperature Range	-65°C to 150°C
Package Thermal Resistance	250°C/W
Power Dissipation (SOT-26, at Ambient Temperature = 85°C)	250mW
Power Dissipation (DIP-8, at Ambient Temperature = 85°C)	650mW
Lead temperature (All PB Free Packages, Soldering, 10sec)	260°C
ESD Voltage Protection, Human Body Model	3.5KV
ESD Voltage Protection, Machine Model	300V
Gate Output Current	300mA

Caution:

Stresses beyond the ratings specified in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.



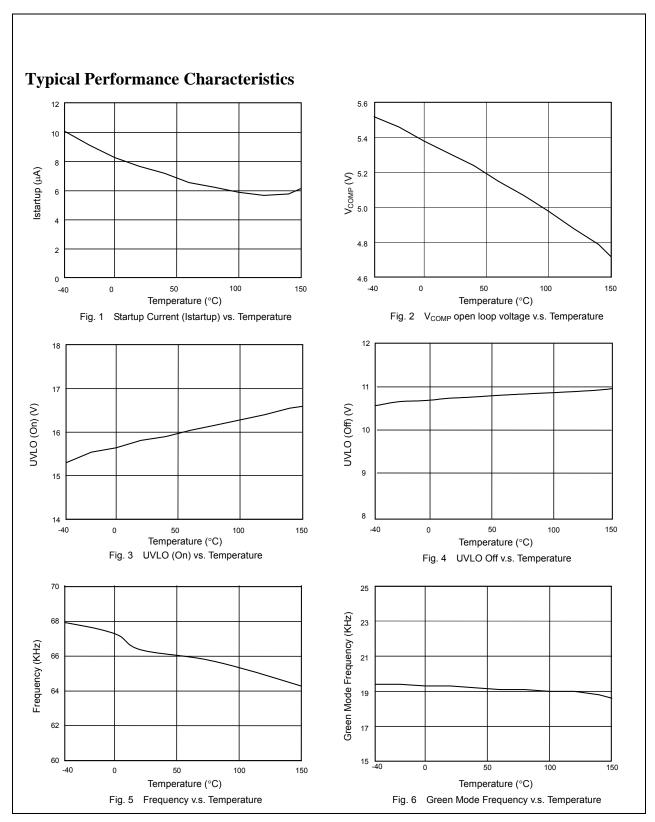
Electrical Characteristics

 $(T_A = +25^{\circ}C \text{ unless otherwise stated}, V_{CC}=15.0V)$

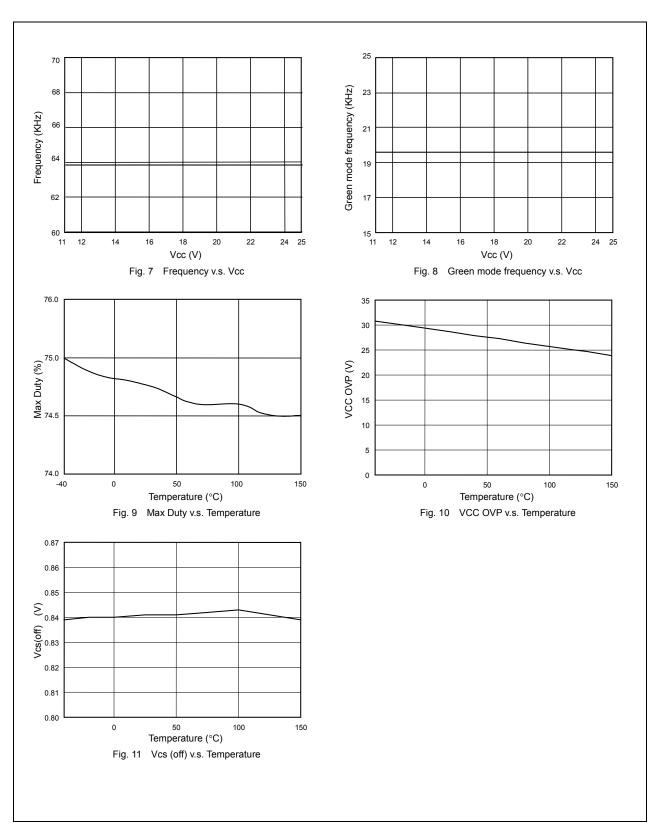
PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
Supply Voltage (Vcc Pin)					
Startup Current			8	20	μА
	V _{COMP} =0V		3	4	mA
Operating Current	V _{COMP} =3V		2		mA
	V _{COMP} =open		0.7		mA
UVLO (off)		9.0	10.0	11.0	V
UVLO (on)		15.0	16.0	17.0	V
OVP Level		27	29	31	V
Voltage Feedback (Comp Pin)	•				_
Short Circuit Current	V _{COMP} =0V		2.2	3.0	mA
Open Loop Voltage	COMP pin open		5.0		V
Green Mode Threshold VCOMP			2.35		V
Current Sensing (CS Pin)	•				
Maximum Input Voltage, Vcs(off)		0.80	0.85	0.90	V
Leading Edge Blanking Time			350		nS
Input impedance			50		ΚΩ
Delay to Output			100		nS
Oscillator (RT pin)					
Frequency	RT=100KΩ	60.0	65.0	70.0	KHz
Green Mode Frequency	Fs=65.0KHz		20		KHz
Temp. Stability	(-40°C ~105°C)			3	%
Voltage Stability	(VCC=11V-25V)			1	%
Gate Drive Output (OUT Pin)					
Output Low Level	VCC=15V, Io=20mA			1	V
Output High Level	VCC=15V, Io=20mA	8			V
Rising Time	Load Capacitance=1000pF		50	200	nS
Falling Time	Load Capacitance=1000pF		30	100	nS

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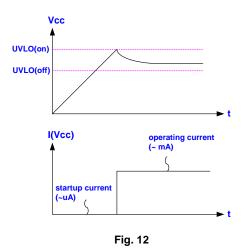
Application Information

Operation Overview

The LD7550B meets the green power requirement and is intended for use in those modern switching power suppliers and switching adaptors, demanding higher power efficiency and power saving. It integrated more functions to reduce the external components counts and the size. Its major features are described as below.

Under Voltage Lockout (UVLO)

An UVLO comparator is implemented to detect the voltage on the VCC pin to ensure the supply voltage is enough to power on the LD7550-B PWM controller and further to drive the power MOSFET. As shown in Fig. 12, a hysteresis is implemented to prevent the shutdown from the voltage dip during startup. The turn-on and turn-off threshold level are set at 16V and 10.0V, respectively.



Startup Current and Startup Circuit

The typical startup circuit to power up the LD7550-B is shown in Fig. 13. During the startup transient, the Vcc is lower than the UVLO threshold thus there is no gate pulse generated from LD7550-B to drive power MOSFET. Therefore, the current through R1 will provide the startup current as well as charge the capacitor C1. Whenever the Vcc voltage is higher enough to power on the LD7550-B and further to deliver the gate drive signal, the supply current is

provided from the auxiliary winding of the transformer. The lower startup current requirement on the PWM controller will help to increase the R1 value and then reduce the power consumption on R1. By using CMOS process and the special circuit design, the maximum startup current of LD7550-B is only $20\mu A$.

If a higher resistance value of the R1 is picked, it will usually take more time to startup. To carefully select the value for R1 and C1 will optimize the power consumption and startup time.

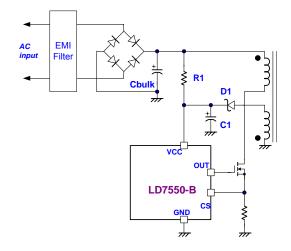


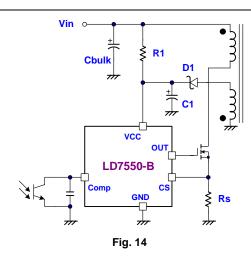
Fig. 13

Current Sensing and Leading-edge Blanking

The typical current mode PWM controller feedbacks both current signal and voltage signal to close the control loop and achieve regulation. As shown in Fig. 14, the LD7550-B detects the primary MOSFET current from the CS pin, which is not only for the peak current mode control but also for the pulse-by-pulse current limit. The maximum voltage threshold of the current sensing pin is set at 0.85V. From above, the MOSFET peak current can be calculated as:

$$I_{PEAK(MAX)} = \frac{0.85V}{R_S}$$





A 350nS leading-edge blanking time is included in the input of CS pin to prevent the false-trigger caused by the current spike and further to eliminate the need of R-C filter which is usually needed in the typical UC384X application (Fig. 15).

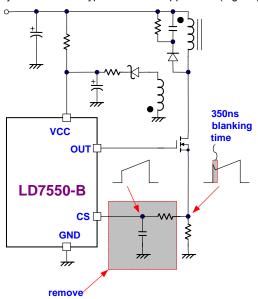


Fig. 15

Output Stage and Maximum Duty-Cycle

An output stage of a CMOS buffer, with typical 300mA driving capability, is incorporated to drive a power MOSFET directly. And the maximum duty-cycle of LD7550-B is limited to 75% to avoid the transformer saturation.

Oscillator and Switching Frequency

Connecting a resistor from RT pin to GND according to the equation can program the normal switching frequency:

$$f_{SW} = \frac{65.0}{RT_{(K\Omega)}} \times 100(KHz)$$

The suggested operating frequency range of LD7550-B is within 50KHz to 130KHz.

Voltage Feedback Loop

The voltage feedback signal is provided from the TL431 in the secondary side through the photo-coupler to the COMP pin of LD7550-B. The input stage of LD7550-B, like the UC384X, is with 2 diodes voltage offset then to feed the voltage divider with 1/3 ratio, that is,

$$V_{+(PWM_{COMPARATOR})} = \frac{1}{3} \times (V_{COMP} - 2V_{F})$$

A pull-high resistor is embedded internally, it can therefore be eliminated from the external circuit.

Internal Slope Compensation

In conventional application, the problem for the stability is a critical issue for current mode controlling, when it operates more than 50% of the duty-cycle. As UC384X, It takes slope compensation from injecting the ramp signal from the RT/CT pin through a coupling capacitor. Well, there is no excess design needed for the LD7550-B since it has integrated this function with it already.

On/Off Control

The LD7550-B can be turnned off by pulling COMP pin lower than 1.2V. The gate output pin of the LD7550-B will be disabled immediately under such condition. The off-mode can be released when the pull-low signal is removed.

Dual-Oscillator Green-Mode Operation

There are many different topologies has been implemented in different chips for the green-mode or power saving requirements such as "burst-mode control", "skipping-cycle mode", "variable off-time control "...etc. The basic operation theory of all these approaches intended to reduce the switching cycles under light-load or no-load condition either





by skipping some switching pulses or reduce the switching frequency.

By using this dual-oscillator control, the green-mode frequency can be well controlled and further to avoid the generation of audible noise.

OVP (Over Voltage Protection) on Vcc

Most of the V_{GS} ratings of the modern power MOSFETs are 30V maximum. To prevent the $_{VGS}$ from the fault condition, LD7550-B also integrates an OVP function on Vcc. Whenever the Vcc voltage is higher than the threshold voltage, the output gate drive circuit will be shut down simultaneously to stop the switching of the power MOSFET.

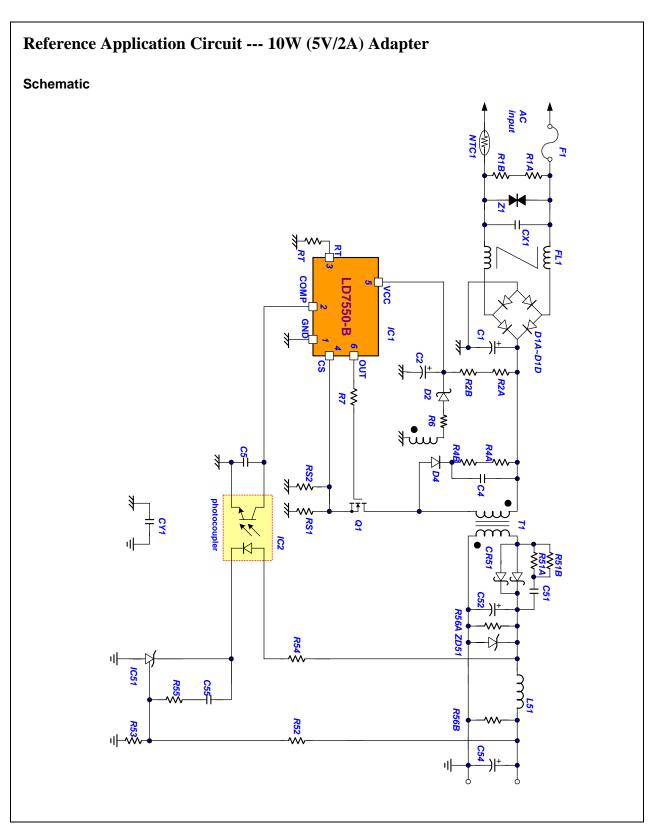
Whenever the Vcc level gets back to lower level, the output will automatically return to the normal operation.

Fault Protection

Several curtail protection features have been integrated in the LD7550-B to protect the power supply or adapter from being damaged. Those damage usually come from open or short condition on the pins of LD7550-B. Under the conditions listed below, the gate output will active off immediately to protect the power circuit ---

- RT pin short to ground
- RT pin floating
- · CS pin floating







Reference Application Circuit --- 10W (5V/2A) Adapter

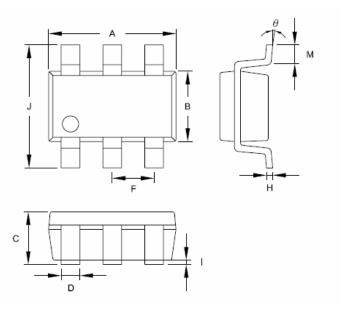
BOM

P/N	Component Value Original		
R1A	N/A		
R1B	N/A		
R2A	750KΩ, 1206		
R2B	750KΩ, 1206		
R4A	39KΩ, 1206		
R4B	39KΩ, 1206		
R6	10Ω, 1206		
R7	10Ω, 1206		
RS1	2.74Ω, 1206, 1%		
RS2	2.74Ω, 1206, 1%		
RT	100ΚΩ, 0805, 1%		
R51A	100Ω, 1206		
R51B	100Ω, 1206		
R52	2.49KΩ, 0805, 1%		
R53	2.49KΩ, 0805, 1%		
R54	220Ω, 0805		
R55	10KΩ, 0805		
R56A	510Ω, 1206		
R56B	N/A		
NTC1	08SP005		
FL1	20mH	UU9.8	
T1	EI-22		
L51	2.7μΗ		

P/N	Component Value	Note
C1	22μF, 400V	L-tec
C2	10μF, 50V	
C4	1000pF, 1000V, 1206	Holystone
C5	0.01μF, 16V, 0805	
C51	1000pF, 50V, 0805	
C52	1000μF, 10V	L-tec
C54	470μF, 10V	L-tec
C55	0.01μF, 16V, 0805	
CX1	0.1μF	X-cap
CY1	2200pF	Y-cap
D1A	1N4007	
D1B	1N4007	
D1C	1N4007	
D1D	1N4007	
D2	PS102R	
D4	1N4007	
Q1	2N60B	600V/2A
CR51	SB540	
ZD51	6V2C	
IC1	LD7550-BBL	SOT-26
IC2	EL817B	
IC51	TL431	1%
F1	250V, 1A	
Z1	N/A	



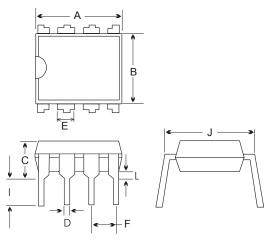
Package Information SOT-26



Symbol	Dimension in Millimeters		Dimensions in Inches	
Symbol	Min	Max	Min	Max
А	2.692	3.099	0.106	0.122
В	1.397	1.803	0.055	0.071
С		1.450		0.058
D	0.300	0.550	0.012	0.022
F	0.838	1.041	0.033	0.041
Н	0.080	0.254	0.003	0.010
ı	0.050	0.150	0.002	0.006
J	2.600	3.000	0.102	0.118
М	0.300	0.600	0.012	0.024
θ	0°	10°	0°	10°



Package Information DIP-8



Symbol	Dimension in Millimeters		Dimensions in Inches	
Symbol	Min	Max	Min	Max
Α	9.017	10.160	0.355	0.400
В	6.096	7.112	0.240	0.280
С		5.334		0.210
D	0.356	0.584	0.014	0.023
Е	1.143	1.778	0.045	0.070
F	2.337	2.743	0.092	0.108
I	2.921	3.556	0.115	0.140
J	7.366	8.255	0.29	0.325
L	0.381		0.015	

Important Notice

Leadtrend Technology Corp. reserves the right to make changes or corrections to its products at any time without notice. Customers should verify the datasheets are current and complete before placing order.



Revision History

Rev.	Date	Change Notice	
00	2/21/2005	Original Specification.	
01	12/22/06	Revision: Block Diagram	
01a	5/17/07	Revision: Marking Description	